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Please cite the original article:

Mäkinen, S. (2020). Discussion on a Device for Measuring 3-Phase Alternating Current at a Distance. In George, Thomas F. (Eds.) *New Insights into Physical Science Vol. 7*. Book Publisher International, 11-19.
<https://doi.org/10.9734/bpi/nips/v7>

Version: Publisher's PDF

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Discussion on a Device for Measuring 3-Phase Alternating Current at a Distance*

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DOI: 10.9734/bpi/nips/v7

ABSTRACT

We have made a preliminary study on a possibility to build a device, which could measure three-phase alternating current at a distance. We wanted to increase the challenge, and decided to consider three-phase alternating current flowing in three wires. In addition, we wanted to consider a meter, which could measure all three currents as functions of time – and so that such a meter could be installed to the transmission line safely without cutting off the current in the line. This report gives a brief look at the theoretical background related with such a device and derives several mathematical result associated with the theory and the structure of the device. Our results suggest that such a device can be built, assuming the distances between the three current-carrying wires and the two sensors of the device are known accurately enough. There are ways to measure these distances, but this article does not concentrate on these.

Keywords: Electromagnetic induction; 3-phase alternating current; ammeter.

1. INTRODUCTION

The transmission line is an important part of the grid. After years of research, power engineers developed a new type of high-level transmission line safe operation monitoring system [1,2]. For energy companies, it is important to know if the transmission lines are working as they should or not. Further, they need this information all the time. Superconducting transmission lines have a tremendous size advantage and lower total electrical losses for high capacity transmission plus a number of technological advantages compared to solutions based on standard conductors [3]. They also want to know about the quality of the electric current flowing in the lines. Power flow control is an electrical device or system that can provide fastacting active or reactive power on transmission lines [4]. Naturally, they have meters to find out the characteristics of the alternating current in their interest, but installation of a meter to a transmission line requires cutting off the current in the line. This is time-consuming and requires more work than the mere installation. As an example, around 21,000 objections were recorded in the case of the Wahle-Mecklar overhead power line (380 kV, 190 km long) in Lower Saxony and Hesse [5], which amounts to about one complaint per 9 m of transmission line.

In our recent paper, we discussed the use of a toroidal coil in measuring alternating current flowing in one wire [6]. We wanted to increase the challenge, and decided to consider three-phase alternating current flowing in three wires. In addition, we wanted to consider a meter, which could measure all three currents as functions of time – and so that such a meter could be installed to the transmission line safely without cutting off the current in the line. We decided to concentrate on the typical 20-kV transmission line, which is shown in Fig. 1.

Since the geometry shown in Fig. 1 is not the only possible one, we need to build our model so that the locations of the three wires are not restricted in any way. This means that we need to introduce a

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system of coordinates fixed to the device, relative to which the locations of the wires are specified. An illustration of the device is added to Fig. 1.



Fig. 1. A typical 20-kV transmission line with three current-conducting wires

2. THE NUMERICAL MODEL

Let us consider three straight and parallel conductors, which carry sinusoidally oscillating currents. We assume that the positions of the wires are not known, and that the phase differences between the three currents are unknown, as well. Further, we assume that the currents may have different amplitudes, and that we do not know these amplitudes beforehand. Thus, the only thing we do know about the case is that the currents are sinusoidal in shape, and that they oscillate at a given frequency. In addition, our analysis is built so that the number of currents is known to be three.

Our task is to find the three currents as functions of time. This means that we have altogether 9 unknowns (if the phase angles φ_1 , φ_2 and φ_3 are ignored in the analysis): the x and the y coordinates of each wire, and the amplitudes of the three currents. For an unequivocal solution, we would thus need at least 9 equations in terms of these 9 variables. Solving such a set of equations seems too time-consuming, and we see that the only practical way to find the currents is to start with more initial information about the system.

In our opinion, the easiest way to obtain more information about the system is to locate the three wires by measuring the distances between these three and some known reference points. With these lengths, one can easily calculate the x and the y coordinates of each wire relative to one of the reference points, and relative to some chosen directions for the x and the y axes.

It appears that only two such points are needed. One point serves as the origin of the Cartesian system of coordinates, and the other shows the direction of, e.g., the x axis. Hence, it would be wise to place two sets of induction coils (or Hall sensors) at two points in the vicinity of the wires, and locate the wires relative to these two.

Using a laser-beam-based device, one can measure the distances between the two reference points – at which the induction coils (or Hall sensors) are placed – and the three wires to a reasonable accuracy. In the following analysis we assume that these lengths have been measured on the installation of the induction coils, and that these lengths are determined accurately enough. We also assume that the distance between the two reference points is known precisely enough.

Relative to the chosen system of coordinates, the measured lengths and the coordinates of the wires are defined according to Fig. 2. In this figure, we have also illustrated the magnetic field vectors

corresponding to the field generated by the current $I_1(t)$ at points P1 and P2, \vec{B}_{11} and \vec{B}_{12} , respectively. Similarly, we have illustrated the magnetic field generated by the second current, $I_2(t)$, at the two points, \vec{B}_{21} and \vec{B}_{22} , and the field produced by the third current, $I_3(t)$, at points P1 and P2, \vec{B}_{31} and \vec{B}_{32} , respectively. The vectors are drawn by assuming, as an example which does not restrict the generality of our analysis, that $t = 0$ and that the currents are given by:

$$\begin{cases} I_1(t) = \hat{I}_1 \cos(\omega t - \varphi_1) \\ I_2(t) = \hat{I}_2 \cos(\omega t - \varphi_2) \\ I_3(t) = \hat{I}_3 \cos(\omega t - \varphi_3) \end{cases} \quad ; \quad \begin{cases} \varphi_1 = 0^\circ \\ \varphi_2 = 120^\circ \\ \varphi_3 = 240^\circ \end{cases}$$

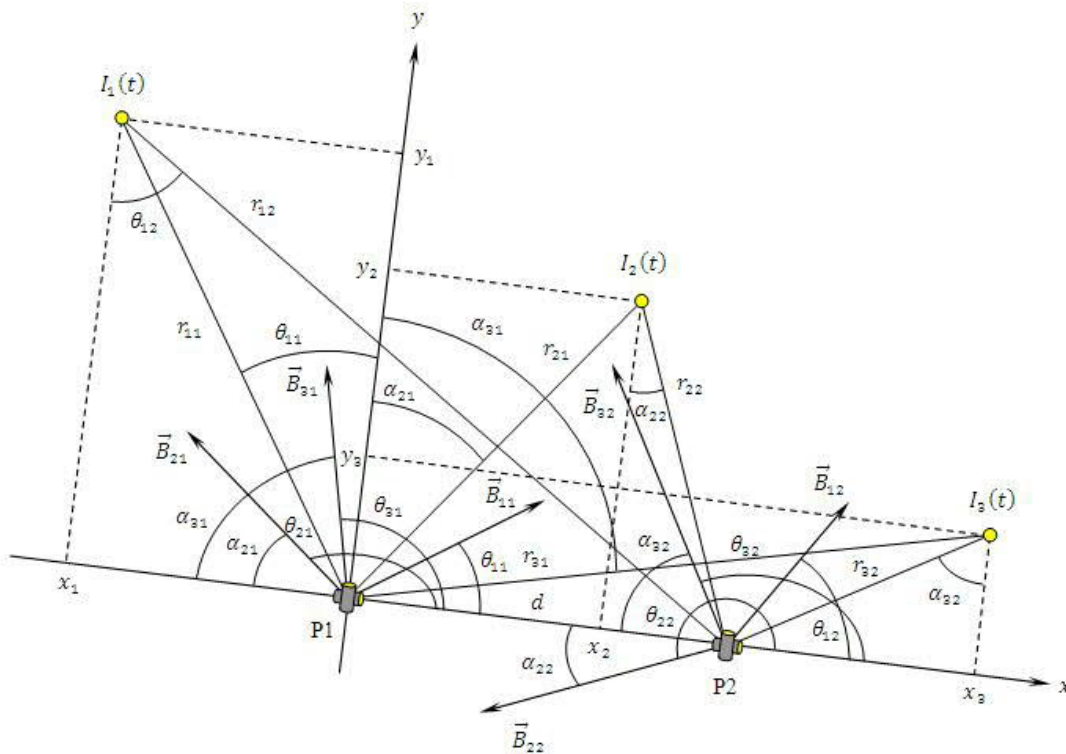


Fig. 2. The two sets of induction coils measure the induced emfs generated by the three sinusoidally oscillating currents

The figure was drawn somewhat slanted to remind the reader about the fact that the x axis does not necessarily have to be horizontal; it only needs to point along a straight line passing through the two reference points. In our analysis, we assume that all the wires are parallel to each other, and that they point along the z axis.

For a case of this kind, we need to use two induction coils at both reference points. One of these is directed along the x axis, while the other points along the y axis.

If the perpendicular distances r_{11} , r_{12} , r_{21} , r_{22} , r_{31} and r_{32} are measured during installation of the induction coils, for instance by using laser beam, one can determine the coordinates of the three wires by applying the Law of Cosines, see Fig. 3:

$$\begin{cases} r_{11}^2 = r_{12}^2 + d^2 - 2r_{12}d \cos \beta_1 \\ r_{21}^2 = r_{22}^2 + d^2 - 2r_{22}d \cos \beta_2 \\ r_{31}^2 = r_{32}^2 + d^2 - 2r_{32}d \cos \beta_3 \end{cases}$$

These yield:

$$\begin{aligned} \beta_1 &= \cos^{-1} \left(\frac{r_{12}^2 + d^2 - r_{11}^2}{2r_{12}d} \right) ; & \begin{cases} x_1 = d - r_{12} \cos \beta_1 \\ y_1 = r_{12} \sin \beta_1 \end{cases} \\ \beta_2 &= \cos^{-1} \left(\frac{r_{22}^2 + d^2 - r_{21}^2}{2r_{22}d} \right) ; & \begin{cases} x_2 = d - r_{22} \cos \beta_2 \\ y_2 = r_{22} \sin \beta_2 \end{cases} \\ \beta_3 &= \cos^{-1} \left(\frac{r_{32}^2 + d^2 - r_{31}^2}{2r_{32}d} \right) ; & \begin{cases} x_3 = d - r_{32} \cos \beta_3 \\ y_3 = r_{32} \sin \beta_3 \end{cases} \end{aligned}$$

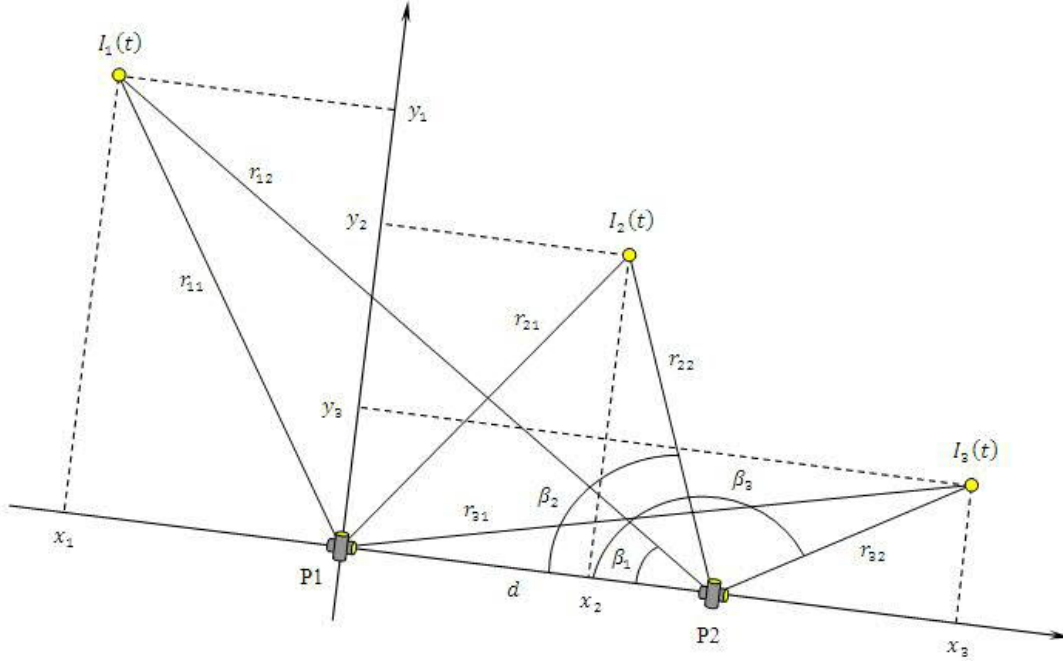


Fig. 3. A simplified illustration of the geometry

The emf induced in the x -directed coil at point P1 can be given in terms of the three currents, the calculated wire coordinates and the angles defined in Fig. 2:

$$\begin{aligned} \varepsilon_{1x}(t) &= -N_{1x} \frac{d}{dt} \{ \vec{A}_{1x} \cdot \vec{B}_{11}(t) + \vec{A}_{1x} \cdot \vec{B}_{21}(t) + \vec{A}_{1x} \cdot \vec{B}_{31}(t) \} \\ &= -N_{1x} \frac{d}{dt} \{ \pi r_{1x}^2 [B_{11}(t) \cos \theta_{11} + B_{21}(t) \cos \theta_{21} + B_{31}(t) \cos \theta_{31}] \} \\ &= -N_{1x} \pi r_{1x}^2 \frac{d}{dt} \left\{ \frac{\mu_0 I_1(t)}{2\pi r_{11}} \cos \theta_{11} + \frac{\mu_0 I_2(t)}{2\pi r_{21}} \cos \theta_{21} + \frac{\mu_0 I_3(t)}{2\pi r_{31}} \cos \theta_{31} \right\} \end{aligned}$$

In this expression, N_{1x} is the number of turns of wire in the induction coil oriented along the x axis and placed at P1; r_{1x} is the radius of this coil. According to Fig. 2, the cosines of the angles in this result can be given as:

$$\cos \theta_{11} = \frac{y_1}{r_{11}} \quad ; \quad \cos \theta_{21} = -\frac{y_2}{r_{21}} \quad ; \quad \cos \theta_{31} = -\frac{y_3}{r_{31}}$$

Therefore, we can give the emf induced in the x -directed coil at P1 as:

$$\varepsilon_{1x}(t) = -\frac{N_{1x} r_{1x}^2 \mu_0}{2} \frac{d}{dt} \left\{ \frac{\hat{I}_1 \cos(\omega t - \varphi_1)}{r_{11}} \frac{y_1}{r_{11}} + \frac{\hat{I}_2 \cos(\omega t - \varphi_2)}{r_{21}} \left(-\frac{y_2}{r_{21}} \right) + \frac{\hat{I}_3 \cos(\omega t - \varphi_3)}{r_{31}} \left(-\frac{y_3}{r_{31}} \right) \right\}$$

$$= \frac{N_{1x}r_{1x}^2\mu_0\omega}{2} \left\{ \frac{y_1}{r_{11}^2} \hat{I}_1 \sin(\omega t - \varphi_1) - \frac{y_2}{r_{21}^2} \hat{I}_2 \sin(\omega t - \varphi_2) - \frac{y_3}{r_{31}^2} \hat{I}_3 \sin(\omega t - \varphi_3) \right\}$$

Similarly, we can show that the emf induced in the x -directed coil at P2 is given by:

$$\varepsilon_{2x}(t) = \frac{N_{2x}r_{2x}^2\mu_0\omega}{2} \left\{ \frac{y_1}{r_{12}^2} \hat{I}_1 \sin(\omega t - \varphi_1) - \frac{y_2}{r_{22}^2} \hat{I}_2 \sin(\omega t - \varphi_2) - \frac{y_3}{r_{32}^2} \hat{I}_3 \sin(\omega t - \varphi_3) \right\}$$

Here, N_{2x} is the number of turns of wire in the second induction coil oriented along the x axis and placed at P2; r_{2x} is the radius of this coil.

We can express these results more simply if we rewrite them in terms of factors k_1, k_2, \dots, k_6 as follows:

$$\varepsilon_{1x}(t) = k_1 \hat{I}_1 \sin(\omega t - \varphi_1) + k_2 \hat{I}_2 \sin(\omega t - \varphi_2) + k_3 \hat{I}_3 \sin(\omega t - \varphi_3) \quad ; \quad \begin{cases} k_1 = \frac{N_{1x}r_{1x}^2\mu_0\omega y_1}{2 r_{11}^2} \\ k_2 = -\frac{N_{1x}r_{1x}^2\mu_0\omega y_2}{2 r_{21}^2} \\ k_3 = -\frac{N_{1x}r_{1x}^2\mu_0\omega y_3}{2 r_{31}^2} \end{cases}$$

$$\varepsilon_{2x}(t) = k_4 \hat{I}_1 \sin(\omega t - \varphi_1) + k_5 \hat{I}_2 \sin(\omega t - \varphi_2) + k_6 \hat{I}_3 \sin(\omega t - \varphi_3) \quad ; \quad \begin{cases} k_4 = \frac{N_{2x}r_{2x}^2\mu_0\omega y_1}{2 r_{12}^2} \\ k_5 = -\frac{N_{2x}r_{2x}^2\mu_0\omega y_2}{2 r_{22}^2} \\ k_6 = -\frac{N_{2x}r_{2x}^2\mu_0\omega y_3}{2 r_{32}^2} \end{cases}$$

The difference in voltage across the two x -directed coils can be given as:

$$\Delta\varepsilon_x(t) = \varepsilon_{2x}(t) - \varepsilon_{1x}(t) = \frac{\mu_0\omega y_1 \hat{I}_1 \sin(\omega t - \varphi_1)}{2} \left\{ \frac{N_{2x}r_{2x}^2}{r_{12}^2} - \frac{N_{1x}r_{1x}^2}{r_{11}^2} \right\} +$$

$$+ \frac{\mu_0\omega y_2 \hat{I}_2 \sin(\omega t - \varphi_2)}{2} \left\{ -\frac{N_{2x}r_{2x}^2}{r_{22}^2} + \frac{N_{1x}r_{1x}^2}{r_{21}^2} \right\} + \frac{\mu_0\omega y_3 \hat{I}_3 \sin(\omega t - \varphi_3)}{2} \left\{ -\frac{N_{2x}r_{2x}^2}{r_{32}^2} + \frac{N_{1x}r_{1x}^2}{r_{31}^2} \right\}$$

If we call the factors that multiply the current amplitudes as k_7, k_8 and k_9 , this result can be given in a shorter form as:

$$\Delta\varepsilon_x(t) = k_7 \hat{I}_1 \sin(\omega t - \varphi_1) + k_8 \hat{I}_2 \sin(\omega t - \varphi_2) + k_9 \hat{I}_3 \sin(\omega t - \varphi_3) \quad ; \quad \begin{cases} k_7 = \frac{\mu_0\omega y_1}{2} \left(\frac{N_{2x}r_{2x}^2}{r_{12}^2} - \frac{N_{1x}r_{1x}^2}{r_{11}^2} \right) \\ k_8 = \frac{\mu_0\omega y_2}{2} \left(\frac{N_{1x}r_{1x}^2}{r_{21}^2} - \frac{N_{2x}r_{2x}^2}{r_{22}^2} \right) \\ k_9 = \frac{\mu_0\omega y_3}{2} \left(\frac{N_{1x}r_{1x}^2}{r_{31}^2} - \frac{N_{2x}r_{2x}^2}{r_{32}^2} \right) \end{cases}$$

Note that the current terms in this result are the actual currents flowing in the three wires, except for the phase difference of $\frac{\pi}{2}$ radians. This means that we can find out the time-dependent currents $I_1(t)$, $I_2(t)$ and $I_3(t)$ – with the phase difference of $\frac{\pi}{2}$ – by solving the following matrix equation:

$$\begin{pmatrix} k_1 & k_2 & k_3 \\ k_4 & k_5 & k_6 \\ k_7 & k_8 & k_9 \end{pmatrix} \begin{pmatrix} I_1(t)' \\ I_2(t)' \\ I_3(t)' \end{pmatrix} = \begin{pmatrix} \varepsilon_{1x}(t) \\ \varepsilon_{2x}(t) \\ \Delta\varepsilon_x(t) \end{pmatrix}$$

Here, the primed symbols refer to the currents, which differ from the real ones by the phase difference of $\frac{\pi}{2}$ rad.

Following a similar approach, one can derive the following result for the y -directed coil at P1:

$$\begin{aligned} \varepsilon_{1y}(t) &= -N_{1y} \frac{d}{dt} \{ \vec{A}_{1y} \cdot \vec{B}_{11}(t) + \vec{A}_{1y} \cdot \vec{B}_{21}(t) + \vec{A}_{1y} \cdot \vec{B}_{31}(t) \} \\ &= -N_{1y} \frac{d}{dt} \left\{ \pi r_{1y}^2 \left[B_{11}(t) \cos\left(\frac{\pi}{2} - \theta_{11}\right) + B_{21}(t) \cos\left(\theta_{21} - \frac{\pi}{2}\right) + B_{31}(t) \cos\left(\theta_{31} - \frac{\pi}{2}\right) \right] \right\} \end{aligned}$$

In this expression, N_{1y} is the number of turns of wire in the induction coil oriented along the y axis and placed at P1; r_{1y} is the radius of this coil. According to Fig. 2 and the fundamental rules of trigonometry, the trigonometric terms in this result can be given as:

$$\begin{aligned} \cos\left(\frac{\pi}{2} - \theta_{11}\right) &= \cos\left(\theta_{11} - \frac{\pi}{2}\right) = \sin \theta_{11} = \frac{x_1}{r_{11}} \\ \cos\left(\theta_{21} - \frac{\pi}{2}\right) &= \sin \theta_{21} = -\sin(2\pi - \theta_{21}) = -\sin \alpha_{21} = -\frac{x_2}{r_{21}} \\ \cos\left(\theta_{31} - \frac{\pi}{2}\right) &= \sin \theta_{31} = -\sin(2\pi - \theta_{31}) = -\sin \alpha_{31} = -\frac{x_3}{r_{31}} \end{aligned}$$

These results yield:

$$\begin{aligned} \varepsilon_{1y}(t) &= -N_{1y} \pi r_{1y}^2 \frac{d}{dt} \left\{ \frac{\mu_0 I_1(t) x_1}{2\pi r_{11}} \frac{1}{r_{11}} - \frac{\mu_0 I_2(t) x_2}{2\pi r_{21}} \frac{1}{r_{21}} - \frac{\mu_0 I_3(t) x_3}{2\pi r_{31}} \frac{1}{r_{31}} \right\} \\ &= -\frac{N_{1y} r_{1y}^2 \mu_0}{2} \frac{d}{dt} \left\{ \frac{x_1}{r_{11}^2} \hat{I}_1 \cos(\omega t - \varphi_1) - \frac{x_2}{r_{21}^2} \hat{I}_2 \cos(\omega t - \varphi_2) - \frac{x_3}{r_{31}^2} \hat{I}_3 \cos(\omega t - \varphi_3) \right\} \\ &= \frac{N_{1y} r_{1y}^2 \mu_0 \omega}{2} \left\{ \frac{x_1}{r_{11}^2} \hat{I}_1 \sin(\omega t - \varphi_1) - \frac{x_2}{r_{21}^2} \hat{I}_2 \sin(\omega t - \varphi_2) - \frac{x_3}{r_{31}^2} \hat{I}_3 \sin(\omega t - \varphi_3) \right\} \end{aligned}$$

Again, we can express this result more simply if we rewrite it in terms of factors p_1 , p_2 and p_3 as follows:

$$\varepsilon_{1y}(t) = p_1 \hat{I}_1 \sin(\omega t - \varphi_1) + p_2 \hat{I}_2 \sin(\omega t - \varphi_2) + p_3 \hat{I}_3 \sin(\omega t - \varphi_3) \quad ; \quad \begin{cases} p_1 = \frac{N_{1y} r_{1y}^2 \mu_0 \omega}{2} \frac{x_1}{r_{11}^2} \\ p_2 = -\frac{N_{1y} r_{1y}^2 \mu_0 \omega}{2} \frac{x_2}{r_{21}^2} \\ p_3 = -\frac{N_{1y} r_{1y}^2 \mu_0 \omega}{2} \frac{x_3}{r_{31}^2} \end{cases}$$

Similarly, we can show that the emf induced in the y -directed coil at P2 is given by:

$$\begin{aligned} \varepsilon_{2y}(t) &= -N_{2y} \frac{d}{dt} \{ \vec{A}_{2y} \cdot \vec{B}_{12}(t) + \vec{A}_{2y} \cdot \vec{B}_{22}(t) + \vec{A}_{2y} \cdot \vec{B}_{32}(t) \} \\ &= -N_{2y} \pi r_{2y}^2 \frac{d}{dt} \left\{ \frac{\mu_0 I_1(t)}{2\pi r_{12}} \cos\left(\frac{\pi}{2} - \theta_{12}\right) + \frac{\mu_0 I_2(t)}{2\pi r_{22}} \cos\left(\theta_{22} - \frac{\pi}{2}\right) + \frac{\mu_0 I_3(t)}{2\pi r_{32}} \cos\left(\theta_{32} - \frac{\pi}{2}\right) \right\} \end{aligned}$$

Here, N_{2y} is the number of turns of wire in the induction coil oriented along the y axis and placed at

P2; r_{2y} is the radius of this coil. According to laws of trigonometry and Fig. 2, the cosines of the angles in this result can be given as:

$$\begin{aligned}\cos\left(\frac{\pi}{2} - \theta_{12}\right) &= \sin \theta_{12} = \frac{d - x_1}{r_{12}} \\ \cos\left(\theta_{22} - \frac{\pi}{2}\right) &= \cos\left(\frac{\pi}{2} - \theta_{22}\right) = \sin \theta_{22} = -\sin(\theta_{22} - \pi) = -\sin \alpha_{22} = -\frac{d - x_2}{r_{22}} \\ \cos\left(\theta_{32} - \frac{\pi}{2}\right) &= \cos\left(\frac{\pi}{2} - \theta_{32}\right) = \sin \theta_{32} = \sin(\pi - \theta_{32}) = \sin \alpha_{32} = \frac{x_3 - d}{r_{32}}\end{aligned}$$

Thus, we can give the emf induced in the y -directed coil at P2 as:

$$\begin{aligned}\varepsilon_{2y}(t) &= -\frac{N_{2y}r_{2y}^2\mu_0}{2} \frac{d}{dt} \left\{ \frac{\hat{I}_1 \cos(\omega t - \varphi_1)}{r_{12}} \frac{d - x_1}{r_{12}} + \frac{\hat{I}_2 \cos(\omega t - \varphi_2)}{r_{22}} \left(-\frac{d - x_2}{r_{22}}\right) \right. \\ &\quad \left. + \frac{\hat{I}_3 \cos(\omega t - \varphi_3)}{r_{32}} \frac{x_3 - d}{r_{32}} \right\} \\ &= -\frac{N_{2y}r_{2y}^2\mu_0\omega}{2} \left\{ -\frac{d - x_1}{r_{12}^2} \hat{I}_1 \sin(\omega t - \varphi_1) + \frac{d - x_2}{r_{22}^2} \hat{I}_2 \sin(\omega t - \varphi_2) + \frac{d - x_3}{r_{32}^2} \hat{I}_3 \sin(\omega t - \varphi_3) \right\} \\ &= \frac{N_{2y}r_{2y}^2\mu_0\omega}{2} \left\{ \frac{d - x_1}{r_{12}^2} \hat{I}_1 \sin(\omega t - \varphi_1) - \frac{d - x_2}{r_{22}^2} \hat{I}_2 \sin(\omega t - \varphi_2) - \frac{d - x_3}{r_{32}^2} \hat{I}_3 \sin(\omega t - \varphi_3) \right\}\end{aligned}$$

Let us rewrite this result in terms of factors p_4 , p_5 and p_6 :

$$\varepsilon_{2y}(t) = p_4 \hat{I}_1 \sin(\omega t - \varphi_1) + p_5 \hat{I}_2 \sin(\omega t - \varphi_2) + p_6 \hat{I}_3 \sin(\omega t - \varphi_3) \quad ; \quad \begin{cases} p_4 = \frac{N_{2y}r_{2y}^2\mu_0\omega}{2} \frac{d - x_1}{r_{12}^2} \\ p_5 = -\frac{N_{2y}r_{2y}^2\mu_0\omega}{2} \frac{d - x_2}{r_{22}^2} \\ p_6 = -\frac{N_{2y}r_{2y}^2\mu_0\omega}{2} \frac{d - x_3}{r_{32}^2} \end{cases}$$

The difference in voltage across the two y -directed coils can be given as:

$$\begin{aligned}\Delta\varepsilon_y(t) &= \varepsilon_{2y}(t) - \varepsilon_{1y}(t) = \frac{\mu_0\omega\hat{I}_1 \sin(\omega t - \varphi_1)}{2} \left\{ N_{2y}r_{2y}^2 \frac{d - x_1}{r_{12}^2} - N_{1y}r_{1y}^2 \frac{x_1}{r_{11}^2} \right\} + \\ &+ \frac{\mu_0\omega\hat{I}_2 \sin(\omega t - \varphi_2)}{2} \left\{ N_{2y}r_{2y}^2 \frac{x_2 - d}{r_{22}^2} + N_{1y}r_{1y}^2 \frac{x_2}{r_{21}^2} \right\} + \frac{\mu_0\omega\hat{I}_3 \sin(\omega t - \varphi_3)}{2} \left\{ N_{2y}r_{2y}^2 \frac{x_3 - d}{r_{32}^2} + N_{1y}r_{1y}^2 \frac{x_3}{r_{31}^2} \right\}\end{aligned}$$

This result can be given in a shorter form as:

$$\begin{aligned}\Delta\varepsilon_y(t) &= p_7 \hat{I}_1 \sin(\omega t - \varphi_1) + p_8 \hat{I}_2 \sin(\omega t - \varphi_2) + p_9 \hat{I}_3 \sin(\omega t - \varphi_3) \quad ; \\ \begin{cases} p_7 = \frac{\mu_0\omega}{2} \left(N_{2y}r_{2y}^2 \frac{d - x_1}{r_{12}^2} - N_{1y}r_{1y}^2 \frac{x_1}{r_{11}^2} \right) \\ p_8 = \frac{\mu_0\omega}{2} \left(N_{2y}r_{2y}^2 \frac{x_2 - d}{r_{22}^2} + N_{1y}r_{1y}^2 \frac{x_2}{r_{21}^2} \right) \\ p_9 = \frac{\mu_0\omega}{2} \left(N_{2y}r_{2y}^2 \frac{x_3 - d}{r_{32}^2} + N_{1y}r_{1y}^2 \frac{x_3}{r_{31}^2} \right) \end{cases}\end{aligned}$$

Note that the current terms in this result are the actual currents flowing in the three wires, except for the phase difference of $\frac{\pi}{2}$ radians.

We can find out the time-dependent currents $I_1(t)$, $I_2(t)$ and $I_3(t)$ – with the phase difference of $\frac{\pi}{2}$ – by solving the following matrix equation:

$$\begin{pmatrix} p_1 & p_2 & p_3 \\ p_4 & p_5 & p_6 \\ p_7 & p_8 & p_9 \end{pmatrix} \begin{pmatrix} I_1(t)' \\ I_2(t)' \\ I_3(t)' \end{pmatrix} = \begin{pmatrix} \varepsilon_{1y}(t) \\ \varepsilon_{2y}(t) \\ \Delta\varepsilon_y(t) \end{pmatrix}$$

Here, the primed symbols refer to the currents, which differ from the real ones by the phase difference of $\frac{\pi}{2}$ rad.

3. APPLYING THE RESULTS

The device described in this article could be used so that both sensors, at P1 and at P2, measure the induced voltage at least once in every millisecond. One measurement sequence could last for about 20 periods of current oscillations, and such sequences could be measured once in every 10 minutes, for example. The collected data could then be transmitted wirelessly to a remote receiver. The received data could be analyzed in the office using software based on the matrix equations derived in this article.

The software could calculate the currents $I_1(t)'$, $I_2(t)'$ and $I_3(t)'$ with both of the given matrix equations. After this, one could compare the results with each other, and if they were close to each other within reasonable accuracy, the final results could be found by calculating the average values of the two. Finally, the actual currents, $I_1(t)$, $I_2(t)$ and $I_3(t)$ would be obtained by shifting the sinusoidal functions of time by $\frac{\pi}{2}$ rad.

The device should be equipped with two additional induction coils along the device's z axis. On installation, one should make sure that the voltage in these coils is zero. This ensures that the two sets of perpendicular coils at P1 and at P2 are in the xy plane perpendicular to the three current-carrying wires.

4. CONCLUSION

Our analysis indicates that it is possible to build a device that can determine three sinusoidally oscillating currents as functions of time at a reasonable distance. The device can be installed safely to the measuring site without turning the power off in the transmission lines. The next logical step in our process would be to build a device explained in this article, and to make test runs with it and real transmission lines.

COMPETING INTERESTS

Author has declared that no competing interests exist.

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He made his doctoral thesis on material sciences to the University of Jyväskylä, Finland. He used positron annihilation technique in studying point defects and dislocations in single crystal semiconductors and insulators. His work included experimental, theoretical and numerical parts, each being equally important to his studies. In addition, he studied effects of particle irradiation on magnetism of powerful permanent magnets. After his academic years, he started working as a principal lecturer of physics at Vaasa University of Applied Sciences, Finland. During the past years, his research work has materialised in several numerical models on ammeters working at a distance, on combustion engines, on extremely high-altitude parachute jumps and on the greenhouse effect. In his latest publication, he has discussed the worst-case scenario related with the use of fossil fuels: How high would the average concentration of atmospheric carbon dioxide increase if all the known resources of natural gas, crude oil and coal were burned out altogether. He has also made digital study material on basic engineering mathematics and physics. The product, VirtuaaliAnkka (CyberDuck), has been used by thousands of Finnish engineering students during the past years, and the product has won a national competition on best practices in engineering study materials. Thus far, he has written 30 scientific research reports, published mostly in international scientific magazines and partly in the form of research reports to companies such as ABB and Wärtsilä Finland.

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This chapter is an extended version of the article published by the same author(s) in the following journal.
Journal of Electromagnetic Analysis and Applications, 7(8): 209-216, 2015.